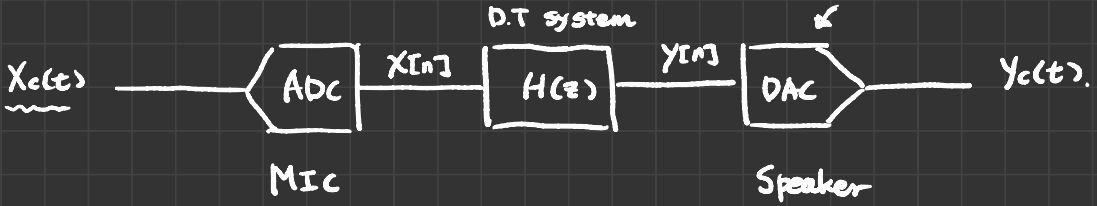
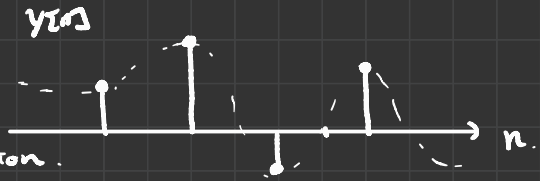


# Reconstruction of CT signals from its samples.



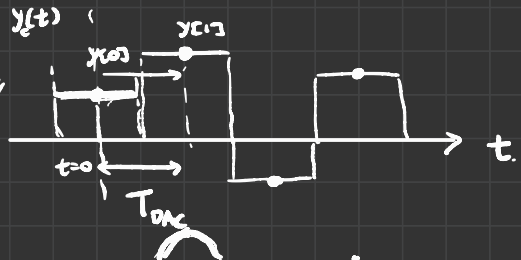
→ Can we have  $x_c(t) = y_c(t)$ ?

DAC: DT → CT  
 $y[n] \xrightarrow{\text{DAC}} y_c(t)$



"  $y_c(t) = \sum_{n=-\infty}^{\infty} g_c(t - nT) \cdot y[n]$  " ← interpolation function.

$= \dots + y[0] \cdot g_c(t) + y[1] \cdot g_c(t - T) + y[2] \cdot g_c(t - 2T) + \dots$



$$\underline{y_c(t)} = \sum_{n=-\infty}^{\infty} g_c(t - nT) \cdot y[n]$$

CTFT  $\updownarrow$

$$\begin{aligned} Y_c(\Omega) &= \int_{-\infty}^{\infty} y_c(t) \cdot e^{-j\Omega t} dt \\ &= \int_{-\infty}^{\infty} \sum_n g_c(t - nT) \cdot \underline{y[n]} \cdot e^{-j\Omega t} dt \\ \hookrightarrow &= \sum_n \underline{y[n]} \int_{-\infty}^{\infty} \underline{g_c(t - nT)} e^{-j\Omega t} dt. \end{aligned}$$

$$\begin{cases} g_c(t) \xleftrightarrow{\text{CTFT}} G_c(\Omega) \\ \rightarrow \underline{g_c(t - nT)} \leftrightarrow \underline{e^{-j\Omega nT} \cdot G_c(\Omega)}. \end{cases}$$

$$= \sum_n y[n] \cdot e^{-j\Omega nT} \cdot \underline{G_c(\Omega)}$$

$$= G_c(\Omega) \cdot \underline{\sum_n y[n] \cdot e^{-j\Omega T n}}$$

$$Y_d(\omega) = \sum_n y[n] e^{-j\omega n}$$

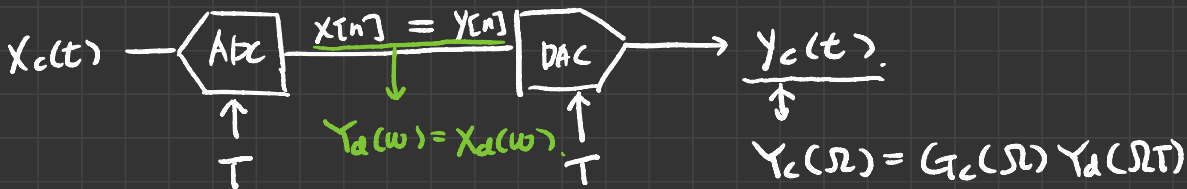
$$= G_c(\Omega) \cdot Y_d(\omega) \Big|_{\omega = \Omega T}$$

$$= \underline{G_c(\Omega) \cdot Y_d(\Omega T)}.$$

$$\underline{y_c(t)} = \sum_{n=-\infty}^{\infty} g_c(t - nT) \cdot y[n]$$

$\updownarrow$  CTFT.

$$Y_c(\Omega) = G_c(\Omega) \cdot Y_d(\Omega T).$$



Assume Nyquist rate sampling at ADC,  $T \cdot \Omega_{\max} < \pi$

$$X_d(\omega) = \frac{1}{T} X_c\left(\frac{\omega}{T}\right) \quad |\omega| \leq \pi$$

We want  $x_c(t) = y_c(t)$ .

Find  $g_c(t)$ .

$$\omega = \Omega \cdot T$$

$$\underline{Y_c(\Omega)} = \underline{G_c(\Omega)} \cdot \underline{Y_d(\Omega T)}$$

$$Y_d(\omega) = X_d(\omega)$$

$$Y_d(\Omega T) = X_d(\Omega T)$$

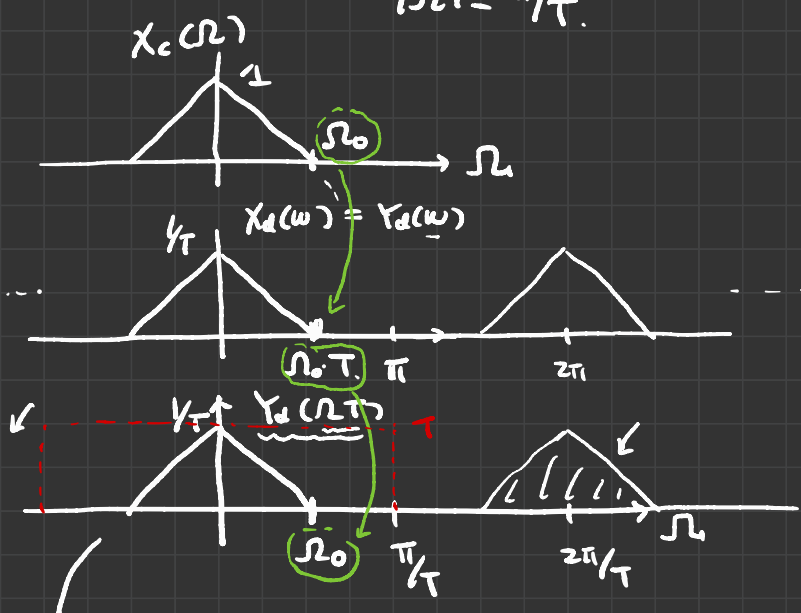
$$= G_c(\Omega) \cdot \frac{1}{T} \cdot X_c(\Omega)$$

$$\therefore X_d(\omega) = \frac{1}{T} X_c\left(\frac{\omega}{T}\right) \quad |\omega| \leq \pi$$

$$Y_d(\Omega T) = \frac{1}{T} X_c(\Omega)$$

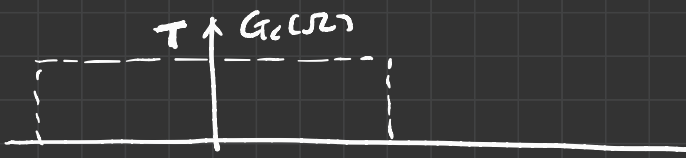
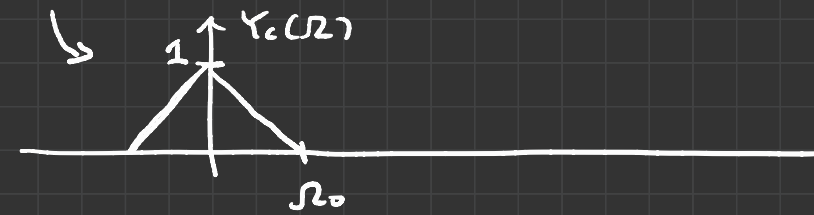
$$|\Omega| \leq \pi/T$$

$$|\Omega| \leq \pi/T$$



$$X_d(\omega) = \frac{1}{T} X_c\left(\frac{\omega}{T}\right)$$

$$Y_d(\Omega T)$$



$\frac{\pi/T}{\text{?}} \rightarrow$  inverse CTFT  
 $g_c(t) = ?$  Sinc function.  
 $= \text{sinc}\left(\frac{\pi t}{T}\right)$

$$G_c(\Omega) = \begin{cases} T & , \quad |\Omega| \leq \pi/T \\ 0 & , \quad \text{otherwise} \end{cases} \quad (\text{Ideal Low-pass Filter})$$

$$g_c(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} G_c(\Omega) e^{j\Omega t} d\Omega \quad (\text{inverse CTFT})$$

$$= \frac{1}{2\pi} \int_{-\pi/T}^{\pi/T} T \cdot \underline{e^{j\Omega t}} d\Omega$$

$$= \frac{T}{2\pi} \left[ \frac{1}{jt} e^{jt\Omega} \right]_{-\pi/T}^{\pi/T}$$

$$= \frac{T}{2\pi} \frac{1}{jt} \cdot \left[ e^{jt\pi/T} - e^{-jt\pi/T} \right]$$

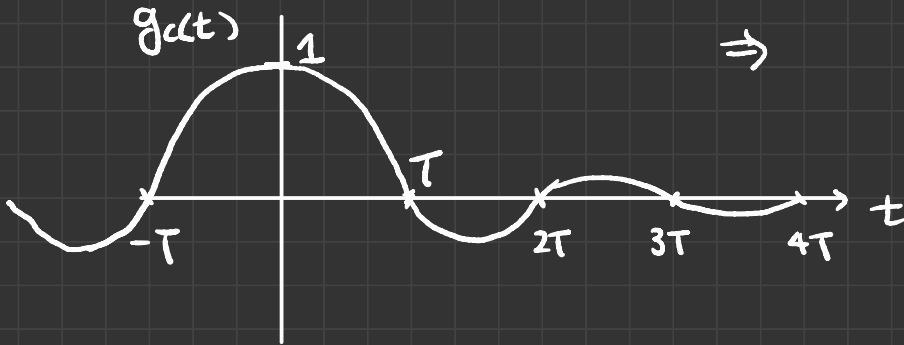
$$\cos t\pi/T + js\sin t\pi/T - (\cos t\pi/T - js\sin t\pi/T)$$

$$2j \sin t\pi/T$$

$$= \frac{T}{2\pi} \cdot \frac{1}{jt} \cdot 2j \sin\left(\frac{\pi}{T} \cdot t\right)$$

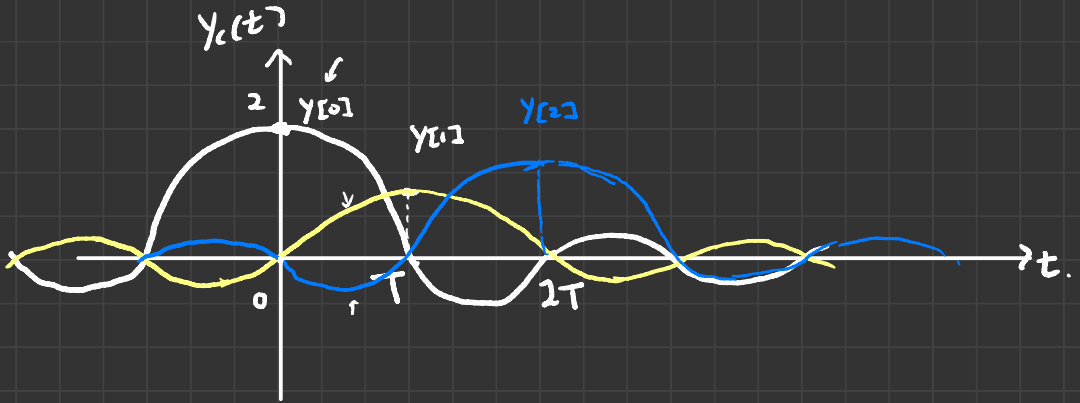
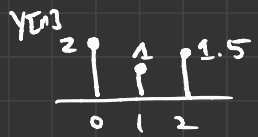
$$= \left( \frac{T}{\pi t} \cdot \sin\left(\frac{\pi}{T} \cdot t\right) \right) \quad \frac{\sin x}{x} = \text{sinc } x$$

$$= \frac{\sin\left(\frac{\pi}{T} \cdot t\right)}{\pi/T \cdot t} = \text{sinc}\left(\frac{\pi t}{T}\right)$$



$g_c(t) = \text{sinc}\left(\frac{\pi t}{T}\right)$  : Ideal DAC  
interpolation  
kernel.

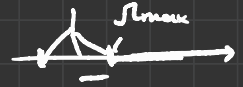
$$y_c(t) = \sum_{n=-\infty}^{\infty} g_c(t - nT) \cdot y[n]$$



## Shannon / Nyquist Sampling Thm

A bandlimited signal  $x_c(t)$

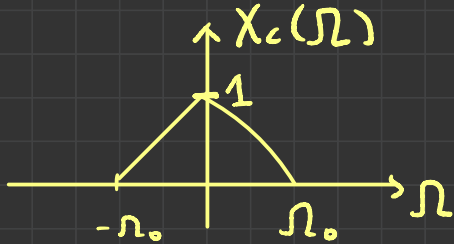
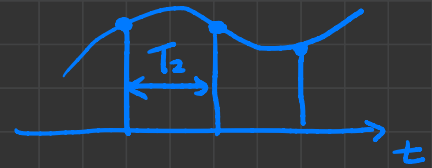
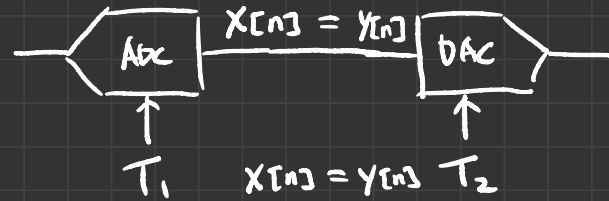
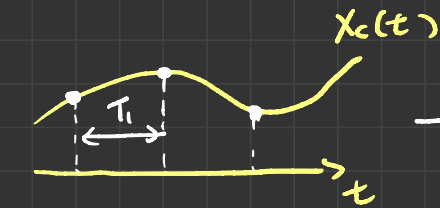
$$(x_c(\Omega) = 0 \text{ for } |\Omega| \geq \Omega_{\max})$$



can be reconstructed perfectly from  
its samples  $x[n]$  taken with sampling rate

$$\frac{1}{T} \geq \frac{\Omega_{\max}}{\pi} = 2 \cdot F_{\max}$$

by the ideal interpolation kernel (sinc).



$X_a(\omega) = Y_a(\omega)$

